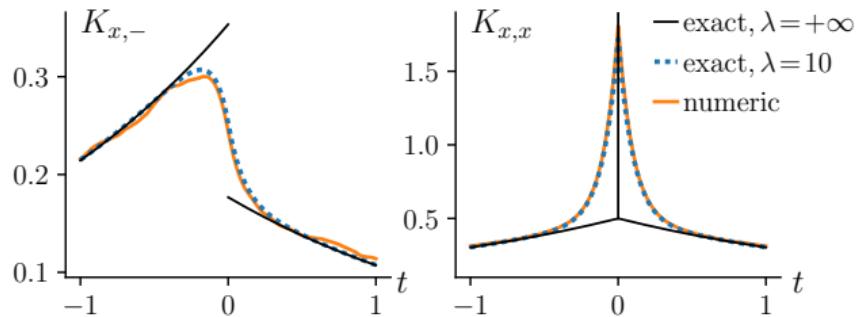


Signal correlation functions for parameter estimation

Antoine Tilloy
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Introduction

There are often two ways to **estimate** things:

- ▶ The **dumb** way: compute theory prediction and fit on simple observables.
- ▶ The **smart** way: reconstruct the optimal Bayesian estimate of the thing from the dataset.



Figure: Chase & Geremia 2009, *Phys. Rev. A* **79**, 022314

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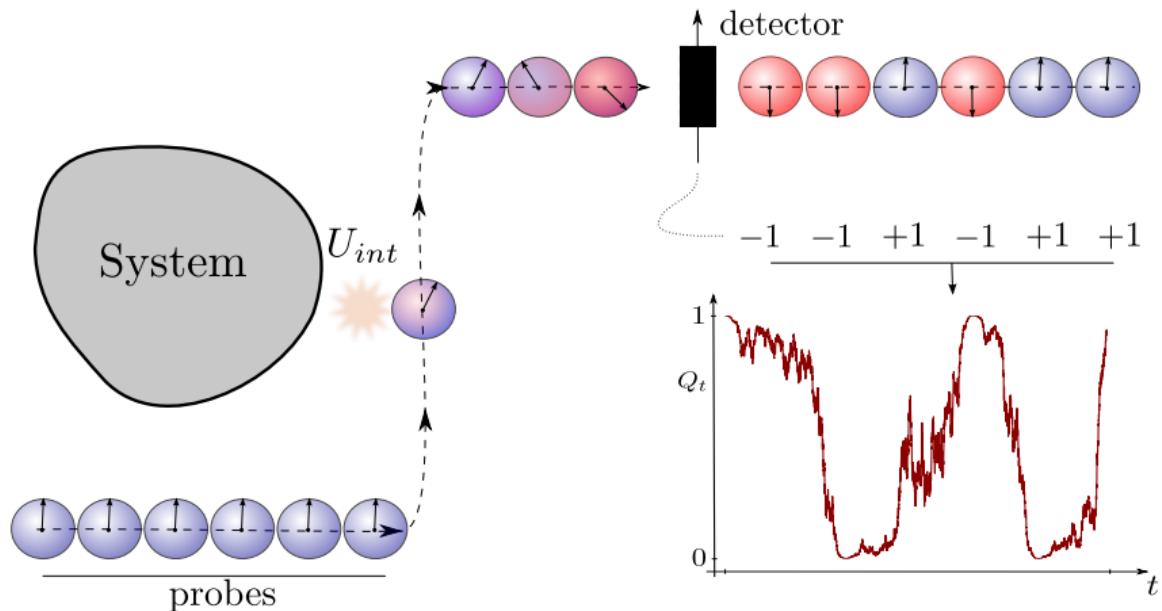
Objective: discuss both approaches in the continuous measurement context

Small novelty: A theoretical formula for the N -point functions makes the **dumb** way quite general

Outline

1. Continuous measurements
2. Smart parameter estimation
3. Dumb parameter estimation
4. Computing correlation function
5. Including imperfections

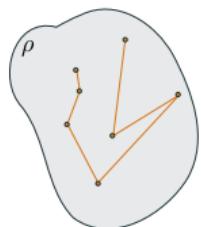
Continuous measurement



Repeated interactions

Discrete quantum trajectories

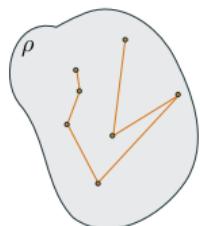
A sequence of $|\psi_n\rangle$ or ρ_n (random) and the corresponding measurement results $\delta_n = \pm 1$.



Repeated interactions

Discrete quantum trajectories

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- ▶ Make the interaction between system and probe smoother

$$U_{\text{int}} = \mathbb{1} + i\sqrt{\varepsilon} \mathcal{O}_{\text{sys}} \otimes K_{\text{probe}}$$

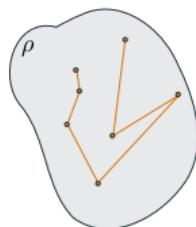
- ▶ Increase the frequency at which probes are sent:

$$\tau \propto \varepsilon$$

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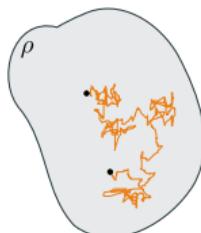
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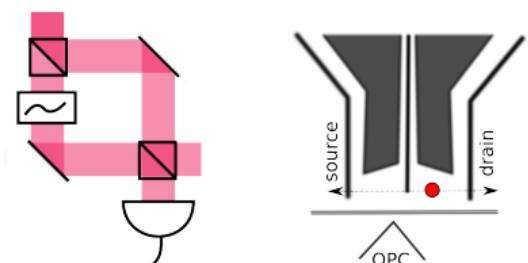
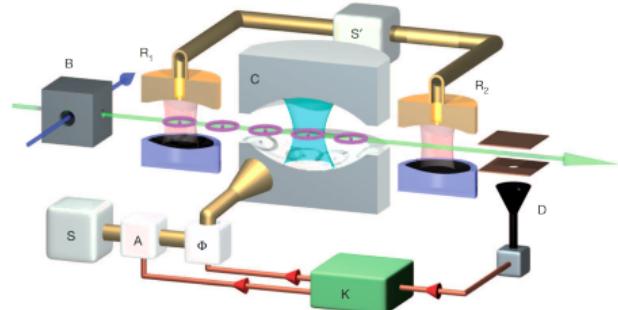
A continuous map $|\Psi_t\rangle$ or ρ_t (random) and the corresponding continuous measurement signal $r_t \propto \sqrt{\varepsilon} \sum_k \delta_k$



⚠ Essentially a central limit theorem result ⚠

In practice

- ▶ Discrete situations “*LKB* style”, with **actual** repeated interactions
- ▶ Almost “true” continuous measurement settings (homodyne detection in quantum optics, quantum point contacts for quantum dots)



Result

Stochastic Master Equation (~ 1987)

Density matrix:

$$d\rho_t = \mathcal{L}(\rho_t) dt + \gamma \mathcal{D}[\mathcal{O}](\rho_t) dt + \sqrt{\gamma\eta} \mathcal{H}[\mathcal{O}](\rho_t) dW_t$$

Signal:

$$I_t = \frac{dr_t}{dt} \text{ with } dr_t = \sqrt{\gamma} \text{tr} \left[(\mathcal{O} + \mathcal{O}^\dagger) \rho_t \right] dt + dW_t$$

with:

- ▶ $\mathcal{D}[\mathcal{O}](\rho) = \mathcal{O}\rho\mathcal{O}^\dagger - \frac{1}{2} (\mathcal{O}^\dagger\mathcal{O}\rho + \rho\mathcal{O}^\dagger\mathcal{O})$
- ▶ $\mathcal{H}[\mathcal{O}](\rho) = \mathcal{O}\rho + \rho\mathcal{O}^\dagger - \text{tr} \left[(\mathcal{O} + \mathcal{O}^\dagger) \rho \right] \rho$
- ▶ $\frac{dW_t}{dt}$ “white noise”



V. Belavkin



A. Barchielli



L. Diósi

Example 0

Situation considered

Pure continuous measurement of a qubit:

- ▶ Qubit $\Rightarrow \mathcal{H} = \mathbb{C}^2$
- ▶ Hence $\rho_t = \begin{pmatrix} p_t & u_t \\ u_t^* & 1 - p_t \end{pmatrix}$
- ▶ Continuous energy measurement: $\mathcal{O} = \sigma_z \propto H$



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Starting point:

$$d\rho_t = \gamma \mathcal{D}[\mathcal{O}](\rho_t) dt + \sqrt{\gamma} \mathcal{H}[\mathcal{O}](\rho_t) dW_t$$



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⇒ Equation for the probability:

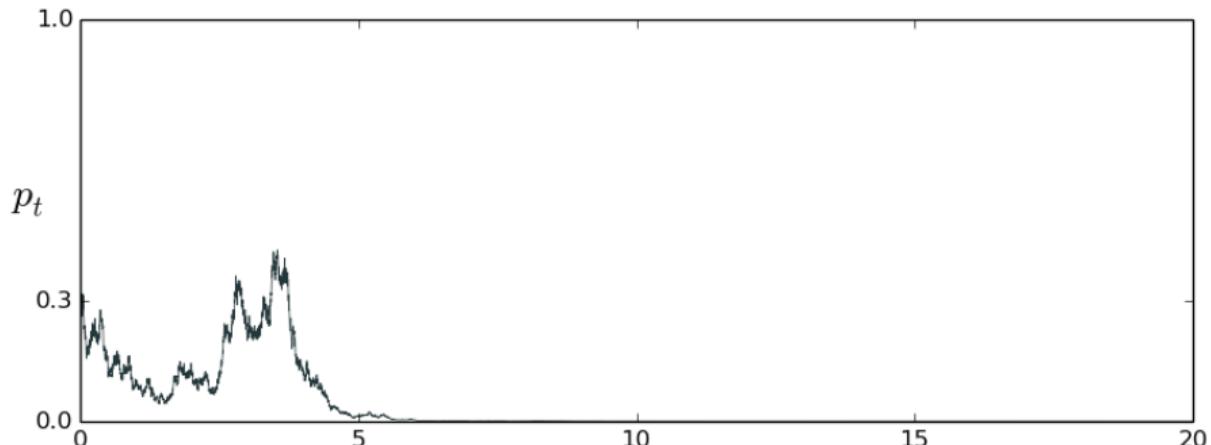
$$dp_t = \sqrt{\gamma} p_t (1 - p_t) dW_t$$

⇒ Equation for the phase:

$$du_t = -\frac{\gamma}{8} u_t dt + \frac{\sqrt{\gamma}}{2} (2p_t - 1) dW_t$$

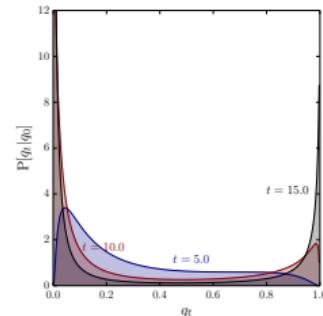
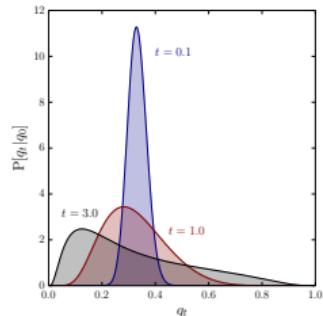
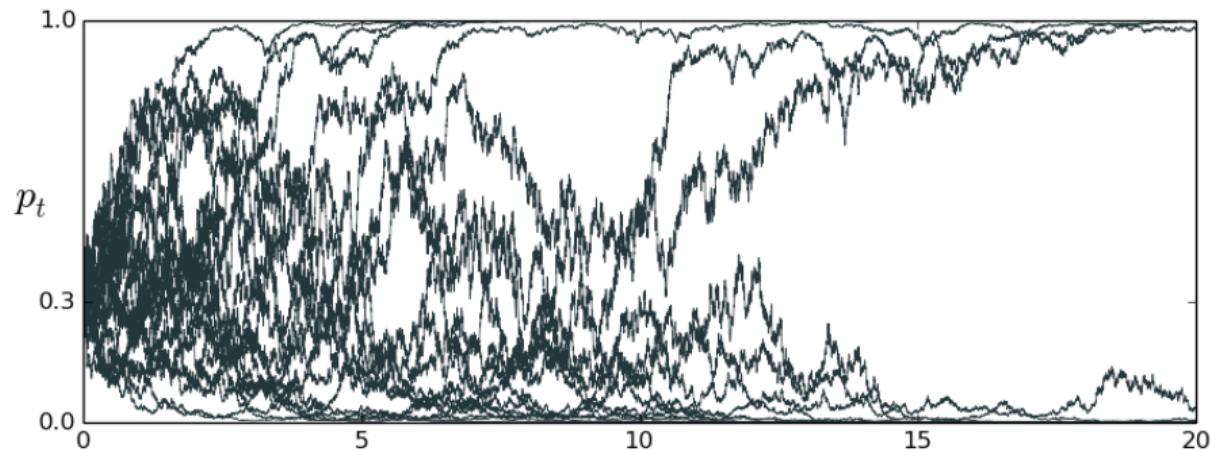
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Smart parameter estimation

We assume there are unknown parameters in the evolution, e.g. ρ evolves with:

$$d\rho_t = -i[H_\xi, \rho_t] dt + \underbrace{\gamma \mathcal{D}[\mathcal{O}](\rho_t) dt + \sqrt{\gamma\eta} \mathcal{H}[\mathcal{O}](\rho_t) dW_t}_{\text{continuous measurement}}$$

with ξ unknown.

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How to estimate ξ from a continuous measurement record I_t ?

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Turn parameter estimation back into a filtering problem.

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Turn parameter estimation back into a filtering problem.

Imagine the unknown parameter is an **internal** degree of freedom:

1. Define $\mathcal{H} = \mathcal{H}_s \otimes \mathcal{H}_{\text{aux}}$ with $\mathcal{H}_{\text{aux}} = \text{span}(|\xi_1\rangle, \dots, |\xi_N\rangle)$
2. Define $\rho_0 = \sum_{k=1}^N \rho_s^k \otimes |\xi_k\rangle\langle\xi_k|$ where $\rho_s^k = \pi_k \rho_s(0)$ where π_k prior for ξ_k
3. Define $H = \sum_{k=1}^N H_{\xi_k} \otimes |\xi_k\rangle\langle\xi_k|$, and $\mathcal{O} \rightarrow \mathcal{O} \otimes \mathbb{1}_{\text{aux}}$
4. Solve the SME: $d\rho_t = -i[H, \rho_t] dt + \gamma \mathcal{D}[\mathcal{O}](\rho_t) dt + \sqrt{\gamma\eta} \mathcal{H}[\mathcal{O}](\rho_t) dW_t$

Diagonality in the auxiliary basis is preserved and generically $\pi_k(t) \underset{t \rightarrow +\infty}{\sim} \delta_{k, k_{\text{real}}}$

See e.g.: Gambetta & Wiseman 2001, *Phys. Rev. A* **64**, 042105 — Chase & Geremia 2009, *Phys. Rev. A* **79**, 022314 — Negretti & Mølmer 2013, *New J. Phys.* **15**, 125002 — Six et al. 2015, 54th IEEE CDC, pp. 7742-7748

Smart parameter estimation: intuition

Why does it work?

Convergence? → Collapse in the space of parameters

Continuous measurement purifies the state enlarged with its “unknown parameter” degree of freedom while preserving the block diagonal structure.

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Why does it work?

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Accuracy? → Hidden Variable

Because we are diagonal in the basis of unknown parameters, the model is mathematically equivalent to a classical filter with an unknown yet predetermined hidden variable ξ .

Easy to extend to a fluctuating unknown parameter $\xi(t)$.

Smart parameter estimation: first subtlety

Discretization of the unknown parameter values $\xi \rightarrow \xi_k$

The auxiliary Hilbert space \mathcal{H}_{aux} needs to be **large** for high accuracy



Smart parameter estimation: first subtlety

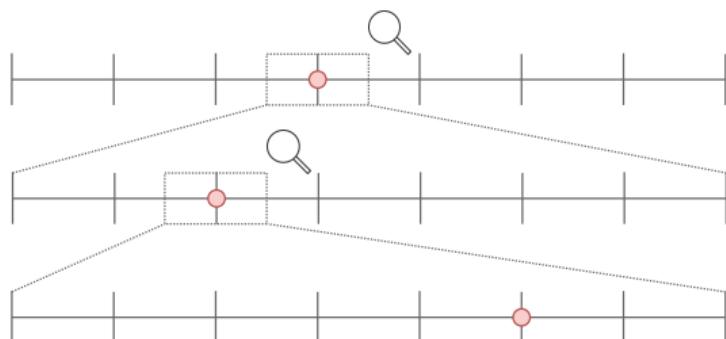
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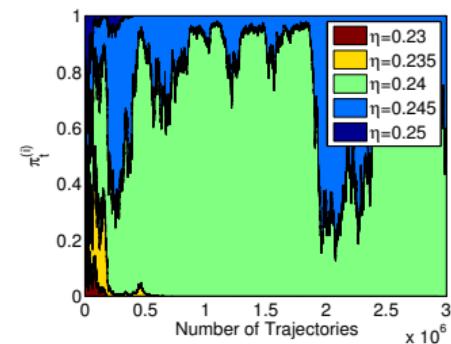
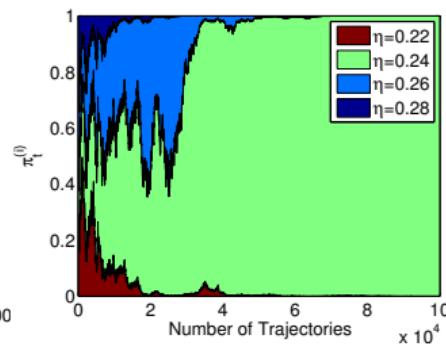
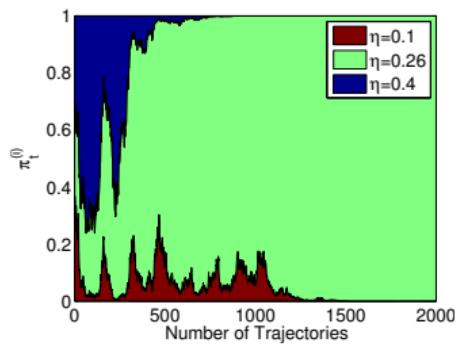
Resampling

Keep \mathcal{H}_{aux} small to find a crude approximation and then refine around this value. Iterate.



Smart parameter estimation: example

Estimation of the efficiency η of the heterodyne detection of a fluorescence:



Smart parameter estimation: difficulties

- a) Main practical difficulty: $\rho_t = f(\rho_0, I)$ where f is a complicated non-linear function of the measured signal.
- b) Further, amplification typically filters the signal:

$$\tilde{I}_t \underset{\text{measured signal}}{=} \int_{-\infty}^t du K(t-u) I_u \underset{\text{true signal}}{=}$$

Solution:

1. **Discretize** at a high frequency
2. **Deconvolute** the filter of the detection chain
3. **Numerically integrate** the discretized SME

But will become untractable for large quantum systems because requires reconstructing ρ .

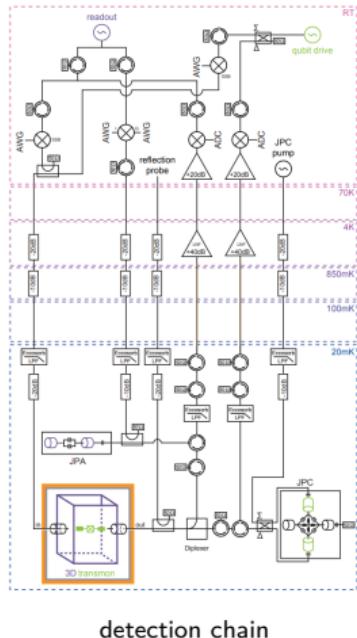


Figure: *Supplementary Material* of Ficheux et al. 2017, arXiv:1711.01208
Smart integration scheme: Rouchon & Ralph 2015, *Phys. Rev. A* **91**, 012118

Dumb parameter estimation

Smart parameter estimation is **mathematically optimal**, but sometimes overkill or hard to implement.

Smart parameter estimation

Reconstruct a non-trivial function of the signal e.g. ρ_t to obtain directly the probability π of the unknown parameter ξ .

Dumb parameter estimation

Compute a trivial function of the signal e.g. $\mathbb{E}[I_{t_1} I_{t_2}]$ and use the **theory** to relate it to system parameters ξ . **Fit** theory on experiment.

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Candidates for “dumb” parameter estimation

→ the signal N -point correlation functions:

$$K_{t_1, \dots, t_N} = \mathbb{E}[I_{t_1} \cdots I_{t_N}]$$

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Advantages

1. Requires only analog manipulations of the signal (delay lines & multipliers)
2. Can include all the imperfections in the theoretical fit.

Computing correlation functions

Setup: Continuous measurement of several operators

$$d\rho = \mathcal{L}(\rho)dt + \sum_{k=1}^n \mathcal{D}[c_k](\rho) dt + \sqrt{\eta_k} \mathcal{H}[c_k](\rho) dW_k,$$

with signal:

$$I_k = \frac{dr_k}{dt} \text{ with } dr_k = \frac{1}{2} \text{tr}[(c_k + c_k^\dagger)\rho] dt + \frac{1}{2\sqrt{\eta_k}} dW_k.$$

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Objective

Compute N -point functions:

$$K_{\ell_1 \ell_2 \dots \ell_N}(t_1, t_2, \dots, t_N) := \mathbb{E} [I_{\ell_1}(t_1) I_{\ell_2}(t_2) \dots I_{\ell_N}(t_N)],$$

as a function of the system dynamics \mathcal{L} , measured operators $\{c_k\}_{k=1}^n$, and efficiencies $\{\eta_k\}_{k=1}^n$.

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Many ways:

- ▶ Quantum noise / continuous matrix product states
- ▶ Discretization

But I take the (possibly pedestrian) **purely stochastic route**.

Computing correlation functions

Idea:

1. Introduce a generating functional $\mathcal{Z}_j(T)$,

$$\mathcal{Z}_j := \mathbb{E} \left[\exp \left(\sum_{k=1}^n \int_0^T j_k(u) I_k(u) \, du \right) \right]$$

2. Solve a differential equation for $\bar{\rho}_j(T)$ such that $\text{tr}[\bar{\rho}_j(T)] = \mathcal{Z}_j(t)$,
3. Carry the functional differentiation explicitly

$$K_{\ell_1 \dots \ell_N}(t_1, \dots, t_N) = \frac{\delta}{\delta j_{\ell_1}(t_1)} \cdots \frac{\delta}{\delta j_{\ell_N}(t_N)} \mathcal{Z}_j \Big|_{j=0}$$

Computing correlation functions

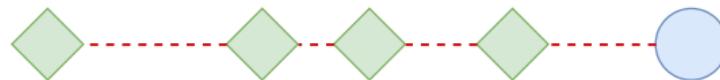
Correlation function – without equal point distribution

For $t_1 < t_2 < \dots < t_N$:

$$K_{\ell_1 \ell_2 \dots \ell_N}(t_1, t_2, \dots, t_N) = \frac{1}{2^N} \times \text{tr} \left[c_{\ell_N}^+ \Phi_{t_N - t_{N-1}} c_{\ell_{N-1}}^+ \dots \Phi_{t_2 - t_1} c_{\ell_1}^+ \Phi_{t_1} \cdot \rho(0) \right],$$

where $\Phi_t = \exp \{t (\mathcal{L} + \sum_{k=1}^n \mathcal{D}[c_k])\}$ and $c^+ \cdot \rho = c\rho + \rho c^\dagger$.

Graphically: initial state – propagators – operator insertions



⚠ Ignores the singular $\delta(t_i - t_j)$ equal point contributions

Close earlier results:

Zoller & Gardiner 1995, *Lecture Notes for the Les Houches Summer School LXIII on Quantum Fluctuations*

Barchielli & Gregoratti 2009, *Quantum trajectories and measurements in continuous time: the diffusive case*

Verstraete & Cirac 2010, *Phys. Rev. Lett.* **104**, 190405

Including imperfections

In practice, signals are smeared (imperfect amplification):

$$I(t) \rightarrow I_\varphi = \int \varphi(u) I(u) du$$

The corresponding correlation function:

$$K_{\ell_1, \dots, \ell_N}(\varphi_1, \dots, \varphi_N) := \mathbb{E} [I_{\ell_1}(\varphi_1) \cdots I_{\ell_N}(\varphi_N)].$$

picks up the equal point δ contributions.

They can be included: **straightforward** albeit slightly cumbersome.

Example

Setup: Qubit without proper dynamics ($\mathcal{L} = 0$), continuously measured with $c_1 = \sqrt{\gamma_x} \sigma_x$ (with associated signal $I_1 := I_x$) and $c_2 = \sqrt{\gamma_-} \sigma_- = \sqrt{\gamma_-} (\sigma_x - i\sigma_y)/2$ (with associated signal $I_2 := I_-$).

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General formula gives e.g. :

$$K_{x,-}(t_1, t_2) = \frac{\sqrt{\gamma_- \gamma_x}}{2} e^{-\gamma_- |t_2 - t_1|/2} \times \left\{ \theta_{t_2 t_1} + \theta_{t_1 t_2} \left[\frac{2\gamma_x}{\gamma_- + 2\gamma_x} - \left(z_0 - \frac{\gamma_-}{\gamma_- + 2\gamma_x} \right) e^{-(\gamma_- + 2\gamma_x)t_1} \right] \right\}$$

with $z_0 = \text{tr}[\sigma_z \rho(0)]$ and $\theta_{t_1 t_2}$ is the Heaviside function

Example

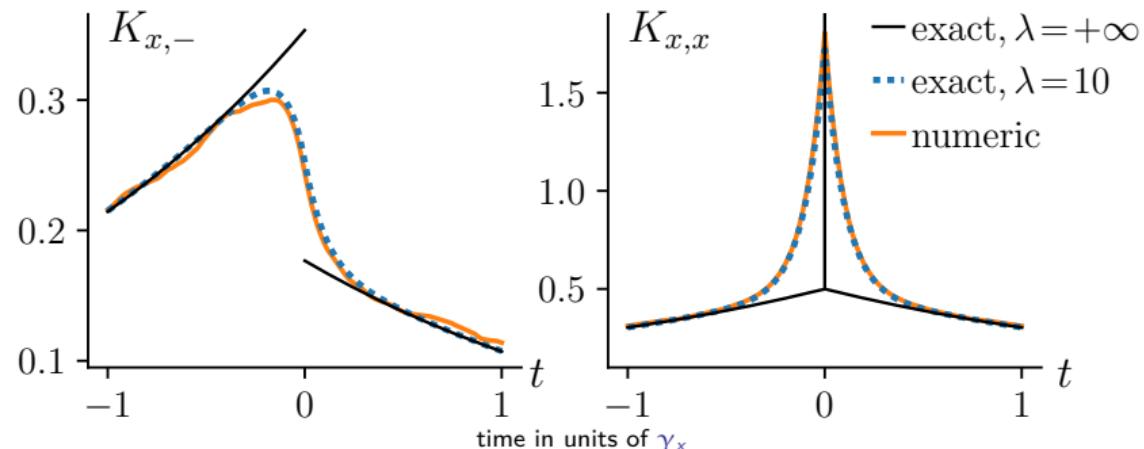
Include a simple imperfection in the detection chain, low pass filter with bandwidth λ :

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- ▶ Picks the otherwise “hidden” equal point singularities

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Link with continuous matrix product states

Continuous matrix product states (cMPS)

A variational class of states for 1 + 1d quantum field theories. Idem continuous measurement with dictionary:

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Continuous measurement translation

All the parameters that can in principle be estimated in the **stationary** state can **generically** be estimated using only 2 and 3 point correlation functions.

Summary

There are **smart** ways and **dumb** ways to estimate system parameters:

- ▶ Smart is optimal, but difficult, and possibly one day untractable.
- ▶ Dumb is suboptimal, but easier to implement

The dumb method relies on the computation of signal N -point functions. The latter can be computed exactly in full generality and including detection imperfections.

Studying and controlling quantum systems without reconstructing their state and focusing only on simple functions of the signal may be easier. Being dumb is the future.

Reference: [arXiv:1712.05725](https://arxiv.org/abs/1712.05725)